Adaptive Inter-chip Interference Cancellation of Multirate Scrambled CDMA Downlink

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Abstract

We propose an adaptive two-stage receiver for a scrambled multirate DS-CDMA downlink. The first stage consists of adaptive FIR equalization generating tentative hard decisions, while the second stage involves adaptive inter-chip interference (ICI) cancellation. Here we detail the adaptive ICI stage, which is based on a low-complexity decision-directed LMS channel identification algorithm; the first stage has been described in a previous publication [1].

1 Introduction

In third generation mobile DS-CDMA systems, downlink multirate symbol streams are multiplexed using orthogonal short codes and then scrambled by a cell-specific long code prior to synchronous transmission. The multipath propagation channel creates interchip interference (ICI) in the received signal, which destroys the orthogonality among user codes. While linear FIR equalization can reduce multi-access interference (MAI) by re-orthogonalizing the chip-rate signal prior to the despreading operation (e.g., [1]), nonlinear processing offers potentially better performance. In [2], tentative chip decisions are used in decision feedback equalization (DFE), but only single-user ICI cancellation is accomplished and the structure is not amenable to adaptation. In [3], a chip-level equalizer output is soft-decoded to obtain estimates of all active symbol streams. These estimates are fed-forward, respread, and used as feedback information for re-processing the received signal via chip-level DFE, cancelling post-cursor ICI.

2 Inter-Chip Interference Cancellation

In this paper we propose an ICI-cancellation stage capable of eliminating both preand post-cursor ICI. Several branches—each corresponding to a different "cursor"—are maximal-ratio combined (MRC) to leverage multipath diversity, as shown in Fig. 1(a). N_{max} is the spreading factor of the lowest-rate user, ν is the system delay through the first stage, $\mathbf{r}_{i-N_{max}-\nu}$ is the delayed-received signal, $\hat{t}_{i-N_{max}-\nu}$ is the detected multiuser chip-rate signal, and $\{\hat{h}_i\}$ is the 1/2-chip spaced channel estimate. (See [4] for a more detailed exposition.) For sparse channels, MRC combines only the largest ICI-cancellation branches. Adaptive LMS channel estimation uses $\hat{t}_{i-N_{max}-\nu}$ as training:

$$\boldsymbol{e}(i) = \hat{\boldsymbol{h}}^{T}(i)\hat{\boldsymbol{t}}(i-N_{max}-\nu) - \boldsymbol{r}_{i-N_{max}-\nu}$$
(1)

$$\hat{\boldsymbol{h}}(i+1) = \hat{\boldsymbol{h}}(i) - \mu \boldsymbol{e}^{T}(i) \otimes \hat{\boldsymbol{t}}^{*}(i-N_{max}-\nu).$$
(2)

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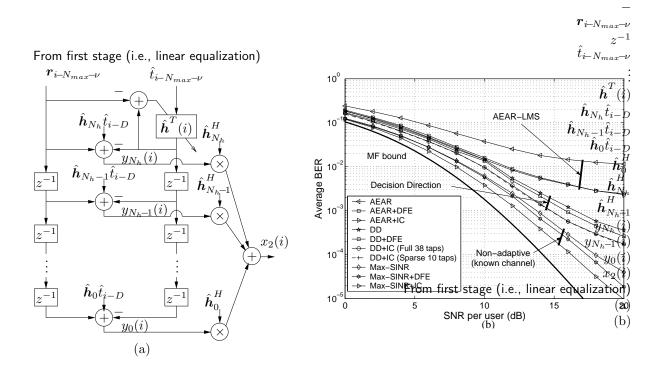


Figure 1: (a): Block diagram of adaptive ICI-canceller. (b): Uncoded BER performance of adaptive and optimal (non-adaptive) algorithms for a half-loaded multirate CDMA downlink with mobile speed of 60 km/hr, 4-chip Rayleigh-fading exponential channel power profile, and 0.22 excess bandwidth. First-stage equalizers span 25 chips (50 taps) and performance is averaged across users. A more detailed description of the simulation can be found in [4].

where \otimes denotes the Kronecker product. Final bit decisions are made from $x_2(i)$.

3 Simulations and Conclusions

Three sets of BER curves are shown in Fig. 1(b) along with the matched filter (MF) bound. Each set shows the result of first-stage linear processing (either AEAR, DD, or Max-SINR) with the possible addition of a second nonlinear processing stage (either DFE or IC). AEAR and DD refer to linear FIR equalizers adapted via pilot-training and decision-direction, respectively, as described in [1], while Max-SINR refers to the SINR-optimal linear FIR equalizer, requiring perfect time-varying channel knowledge. We attribute the superiority of the adaptive ICI canceller "IC," relative to the adaptive DFE (e.g., [3]), to its simultaneous pre- and post-cursor ICI cancellation ability. Furthermore, ICI cancellation with sparse MRC (10 largest taps) performs on par with full MRC (38 taps). Both IC and DD outperform linear processing alone.

References

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